

# Digital Control

## Lecture 3



# Outline

- 1 Digital Controller Design
- 2 Lead Compensator for Antenna - Design Example



# What have we talked about in MIM2?

- Sampling rate selection
- Equivalents between continuous & digital Systems

# Sampling theorem

## Theoretical lower limit

$$\frac{\omega_s}{\omega_b} > 2$$

$\omega_s$ : sampling frequency

$\omega_b$ : required closed-loop bandwidth

# Smooth response

## Practical limits

$$20 < \frac{\omega_s}{\omega_b} < 40$$

$\omega_s$ : sampling frequency

$\omega_b$ : required closed-loop bandwidth

# Requirements to prefilter

Breakpoint of prefilter,  $\omega_c$

$$\omega_b < \omega_c < \frac{\omega_s}{2}$$

$\omega_s$ : sampling frequency

$\omega_b$ : required closed-loop bandwidth

$\omega_c$ : breakpoint of filter

**Prefilter should filter lowest noise frequencies while not disturbing highest system frequencies!**



## Accommodating for sampling effect delay

### Approximation of delay

$$e^{-sT/2} \approx \frac{2/T}{s + 2/T}$$

### Accommodation

Investigate effect of half-sample delay on continuous system before digitizing:

- Root locus wrt.  $T$
- Analyze using frequency based methods
  - Effect on phase margin/damping



# Digital equivalents of continuous-time specifications

## Transient response

- Overshoot/damping,  $M_p, \zeta$
- Rise time,  $\omega_n$
- Settling time,  $\omega_n, \zeta, \sigma$

## Steady-state response

- Steady-state:  $z \rightarrow 1, (\lim_{k \rightarrow \infty} f(k) = \lim_{z \rightarrow 1} (z - 1)F(z))$
- System type: number of pure integrators in open-loop, ( $z = 1$ )
- System input
  - Step,  $\frac{z}{z-1}$
  - Ramp,  $\frac{Tz}{(z-1)^2}$
  - Parabola,  $\frac{T^2}{2} \frac{z(z+1)}{(z-1)^3}$



# Outline

- 1 Digital Controller Design
  - Emulation Method for Digital Control
  
- 2 Lead Compensator for Antenna - Design Example
  - Effect of Sample Times
  - Accommodation for Sampling Delay
  - Effect of Sampling Method





# Digital Controller Design

Digital controller can be obtained using:

- Emulation, which finds the discrete equivalent of a continuous controller
- Direct discrete design (next lecture)



# Frequency Issues

## Continuous Systems

For a minimum-phase transfer function, the phase is uniquely determined by the magnitude curve:

$$\angle G(j\omega) \approx n \times 90^\circ$$

where  $n$  is the slope of  $G(j\omega)$  in units of decade of amplitude

## Discrete Systems

The amplitude and phase relationship is lost!

The prediction of stability from the amplitude curve alone for minimum-phase systems is lost

It is typically necessary to determine both magnitude and phase for discrete systems



# Emulation Method

- 1 A *continuous* controller is designed
- 2 Sample time is selected
- 3 Discrete equivalent is computed
- 4 Evaluation of design



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## Case Study: Antenna Control

General System Model:

$$J\ddot{\theta} + B\dot{\theta} = T_c + T_d$$

Discarding the disturbances  $T_d$  gives the transfer function:

$$\frac{\Theta(s)}{U(s)} = \frac{1}{s\left(\frac{s}{a} + 1\right)}$$

where  $a = \frac{B}{J} = 0.1$  and  $u(t) = \frac{T_c(t)}{B}$ .

### Design Specifications:

- Overshoot to a step input less than 16% ( $PM \approx 55$ )
- Settling time to 1% in less than 10s
- Tracking error to ramp of slope  $0.01 \frac{\text{rad}}{\text{sec}}$  less than 0.01rad
- Sampling time to give at least 10 samples in a rise-time



# Lead Compensator Design for Antenna (FC pp. 375)

## Step 1

Design the low frequency gain  $K$  with respect to the steady-state error specification

Antenna system case:  $K = 1$

## Step 2

Determine the needed phase lead

```
sys=tf(1,[10 1 0]);  
margin(sys)
```

PM=18 at  $\omega = 0.308$



# Lead Compensator Design for Antenna (FC pp. 375)

## Step 3

Using lead contribution of  $\phi_{max} = 45$  should result in PM=63 which is 8 more than needed.

## Step 4

Determine:

$$\alpha = \frac{1 - \sin \phi_{max}}{1 + \sin \phi_{max}} = \frac{1 - \sin 45}{1 + \sin 45} = 0.1716$$

## Step 5

$$T = \frac{1}{\omega_{max} \sqrt{\alpha}} = \frac{1}{\frac{\omega_n}{2} \sqrt{\alpha}} = \frac{2}{0.92 \sqrt{\alpha}} = 5.248$$

Giving a zero in  $s = -\frac{1}{T} = -0.19$  and a pole in  $s = -\frac{1}{\alpha T} = -1.11$ .



## Lead Compensator Design for Antenna (FC pp. 375)

## Step 6

Draw the compensated frequency response, check PM  
Using the formulation:

$$D(s) = \frac{Ts + 1}{\alpha Ts + 1}$$

we use:

```
sysD=tf([5.3 1],[0.9 1])  
sysC=sys*sysD  
margin(sysC)  
step(feedback(sysC,1))
```





# Lead Compensator Design for Antenna (FC pp. 375)

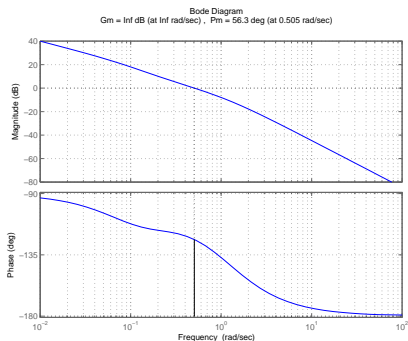


Figure: Frequency response

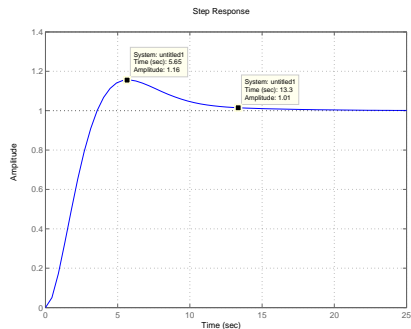


Figure: Step response

# Lead Compensator Design for Antenna (FC pp. 375)

## Step 7

**Step 7:** Iterate on the design until all specifications are met

```
sysD=tf([10 1],[1 1])  
sysC=sys*sysD  
margin(sysC)  
sysCL=feedback(sysC,1)  
step(sysCL)
```



# Lead Compensator Design for Antenna (FC pp. 375)

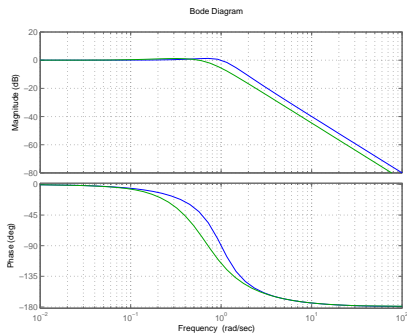


Figure: Frequency response

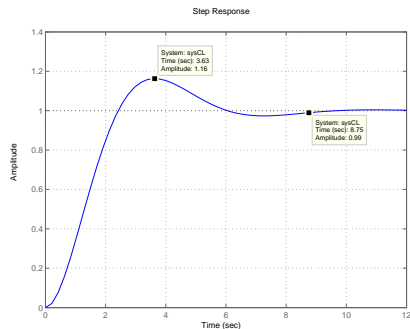


Figure: Step response

# Digital Lead Compensator for Antenna - Fast Sampling

## Continuous lead controller

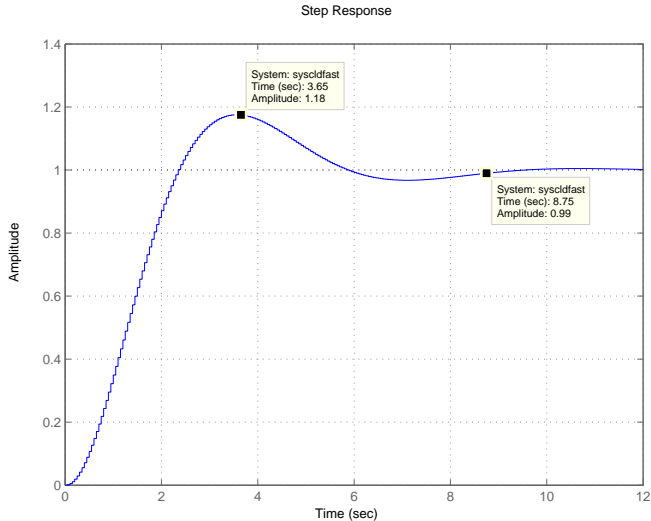
$$D(s) = \frac{10s + 1}{s + 1}$$

## Digitization - Fast Sample Rate

```
sysc=tf(1,[10 1 0]);  
lead=tf([10 1],[1 1]);  
syslead=sysc*lead;  
Ts=1/20;  
leadd1=c2d(lead,Ts,'zoh');  
sysd=c2d(sysc,Ts,'zoh');  
syscld=feedback(sysd*leadd1,1);  
step(syscld)
```



# Digital Lead Compensator for Antenna - Fast Sampling



# Digital Lead Compensator for Antenna - Slow Sampling

## Continuous lead controller

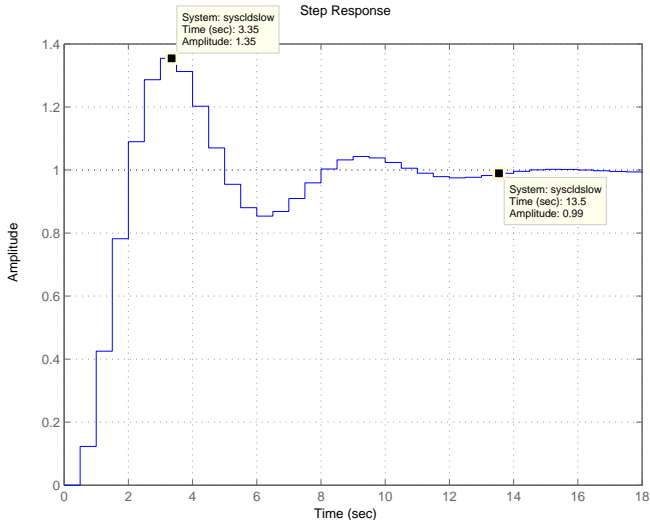
$$D(s) = \frac{10s + 1}{s + 1}$$

## Digitization - Slow Sample Rate

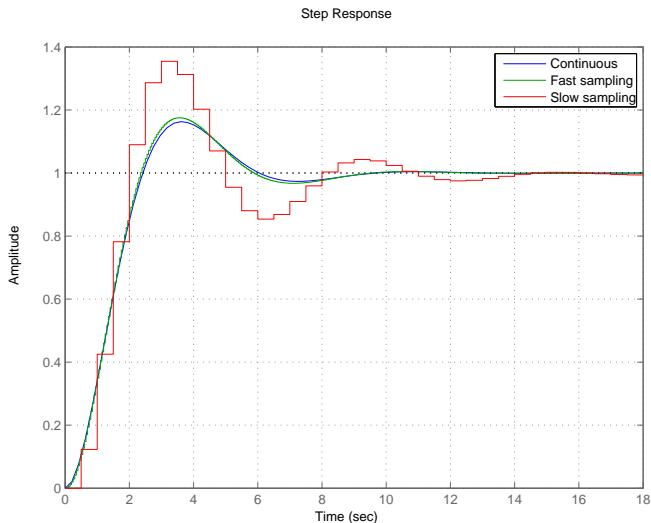
```
sysc=tf(1,[10 1 0]);  
lead=tf([10 1],[1 1]);  
syslead=sysc*lead;  
Ts=1/2;  
leadd1=c2d(lead,Ts,'zoh');  
sysd=c2d(sysc,Ts,'zoh');  
syscld=feedback(sysd*leadd1,1);  
step(syscld)
```



# Digital Lead Compensator for Antenna - Slow Sampling

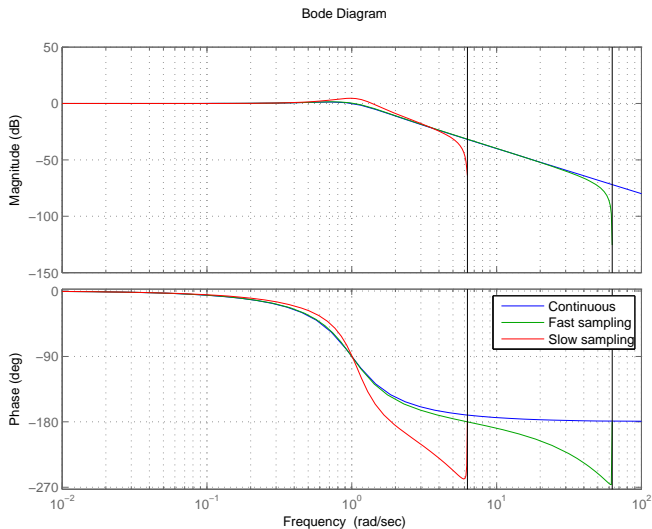


# Effect of Sample Time on Step Response

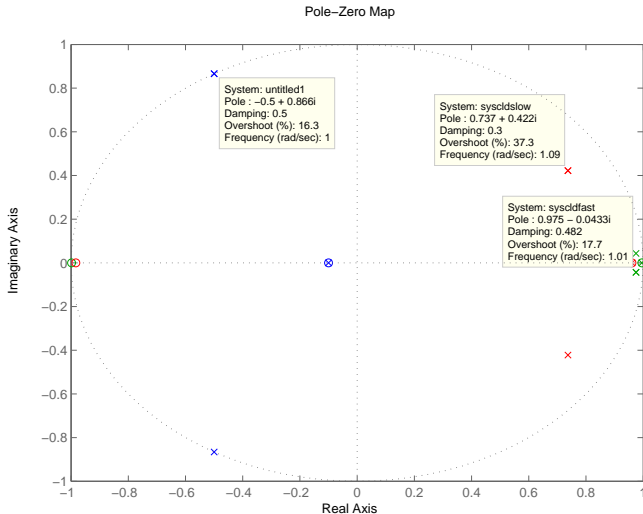




# Effect of Sample Time on Frequency Response



## Effect of Sample Time on Pole Locations



# Incorporating Sampling Delay in System

## Continuous System

$$G(s) = \frac{1}{s(10s + 1)}$$

## Continuous System with Delay

$$G_d(s) = \frac{2/T}{s + 2/T} \frac{1}{s(10s + 1)}$$



## Lead Compensator for System using Slow Sampling Rate

Inserting  $T = 1/2$ 

$$\begin{aligned} G_d(s) &= \frac{2/T}{s + 2/T} \frac{1}{s(10s + 1)} \\ &= \frac{4}{s(s + 4)(10s + 1)} \end{aligned}$$



# Lead Compensator Design for Antenna

## Step 1

Design the low frequency gain  $K$  with respect to the steady-state error specification

Steady-state unchanged from original system:  $K = 1$

## Step 2

Determine the needed phase lead

```
sys=tf(1,[10 41 4 0]);  
margin(sys)
```

PM=14 at  $\omega = 0.308$



# Lead Compensator Design for Antenna

## Step 3

Using lead contribution of  $\phi_{max} = 50$  should result in PM=64 which is 9 more than needed.

## Step 4

Determine:

$$\alpha = \frac{1 - \sin \phi_{max}}{1 + \sin \phi_{max}} = \frac{1 - \sin 50}{1 + \sin 50} = 0.1325$$

## Step 5

$$T = \frac{1}{\omega_{max} \sqrt{\alpha}} = \frac{1}{0.4 \sqrt{(0.1325)}} = 6.869$$

Giving a zero in  $s = -\frac{1}{T} = -0.1456$  and a pole in  $s = -\frac{1}{\alpha T} = -1.099$ .



# Lead Compensator Design for Antenna

## Step 6

Draw the compensated frequency response, check PM  
Using the formulation:

$$D(s) = \frac{Ts + 1}{\alpha Ts + 1}$$

we use:

```
sysD=tf([6.9 1],[0.9 1])  
sysC=sys*sysD  
margin(sysC)  
step(feedback(sysC,1))
```



# Lead Compensator Design for Antenna

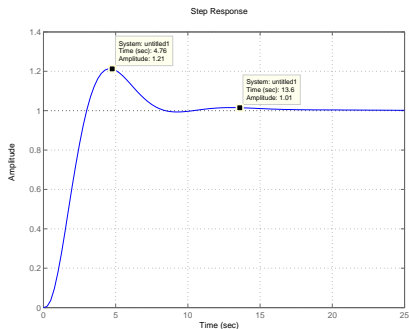
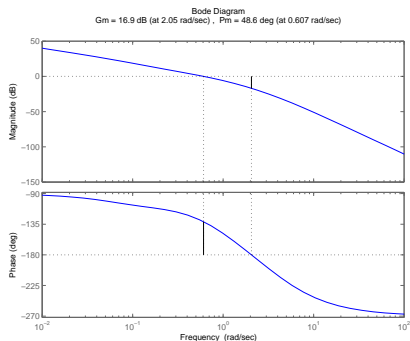


Figure: Frequency response

Figure: Step response





# Lead Compensator Design for Antenna

## Step 7

**Step 7:** Iterate on the design until all specifications are met

```
sysD=tf([7.5 1],[0.68 1])  
sysC=sys*sysD  
margin(sysC)  
sysCL=feedback(sysC,1)  
step(sysCL)
```



# Lead Compensator Design for Antenna

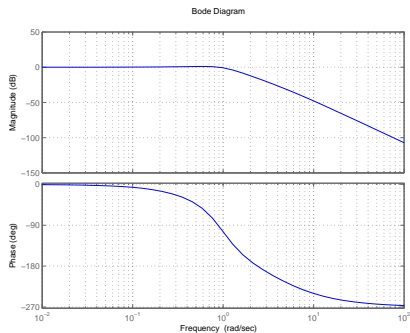


Figure: Frequency response

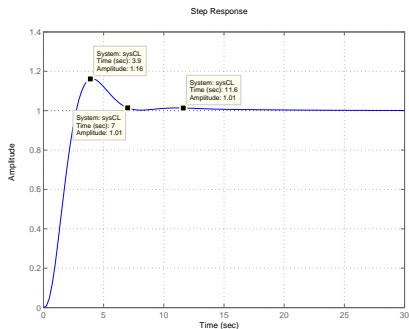


Figure: Step response

# Digital Lead Compensator for Antenna - Slow Sampling

## Continuous lead controller

$$D(s) = \frac{7.5s + 1}{0.68s + 1}$$

## Digitization - Slow Sample Rate

```
sysc=tf(1,[10 1 0]);  
lead=tf([7.5 1],[0.68 1]);  
syslead=sysc*lead;  
Ts=1/2;  
leadd1=c2d(lead,Ts,'zoh');  
sysd=c2d(sysc,Ts,'zoh');  
syscld=feedback(sysd*leadd1,1);  
step(syscld)
```



# Digital Lead Compensator for Antenna - Comparison

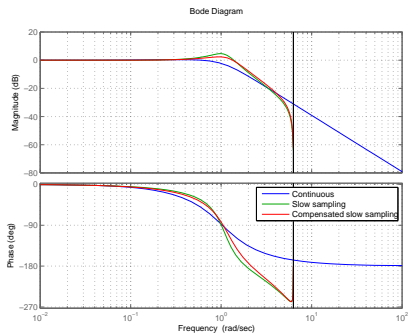


Figure: Frequency response

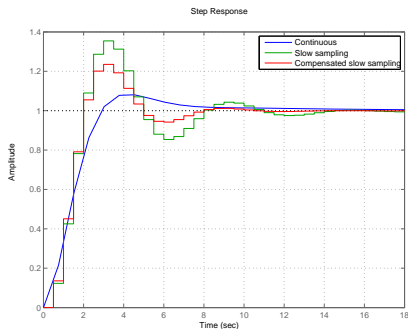


Figure: Step response

# Discretization in MATLAB

## MATLAB

```
sysd=c2d(sys,Ts,method)
```

method:

- 'zoh': Zero order hold
- 'foh': First order hold (academic)
- 'tustin': Bilinear approximation (trapezoidal)
- 'prewarp': Tustin with a specific frequency used for prewarp
- 'matched': Matching continuous poles with discrete



## Discretization of Lead Compensator - Fast Sample Rate

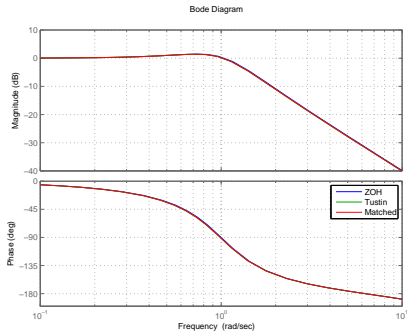


Figure: Frequency response

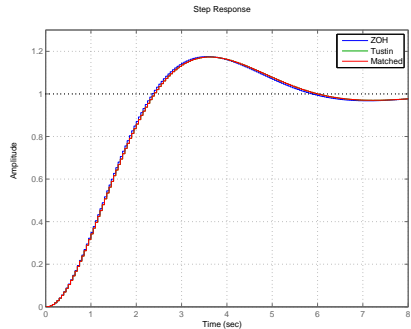
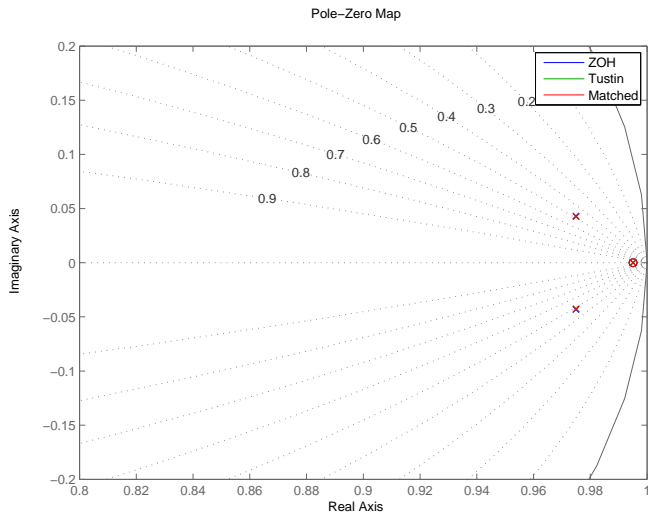


Figure: Step response

# Discretization of Lead Compensator - Fast Sample Rate



# Discretization of Lead Compensator - Slow Sample Rate

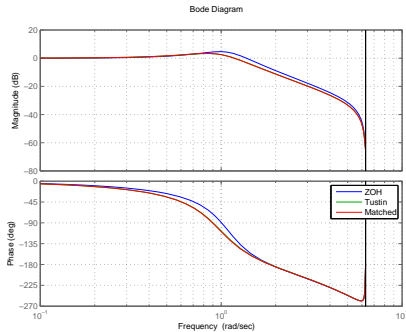


Figure: Frequency response

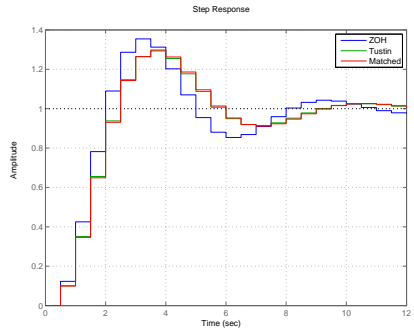
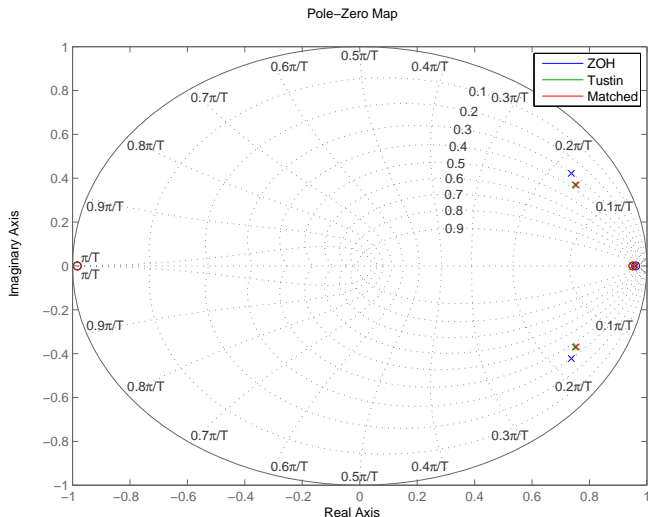


Figure: Step response



# Discretization of Lead Compensator - Slow Sample Rate



## Some important things to remember

### Discretization of compensator

Use the method suited for implementation in the system

### Discrete equivalent of plant

- Use method corresponding to implementation (usually ZOH)
- Simulink can combine discrete compensator with continuous plant (digitization of plant not necessary)



## Book: Digital Control

- Problem 7.4
- Problem 7.5
- Problem 7.7

